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Trajectory Tracking for Chaos Synchronization via PI Control Law between Roosler-Chen
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Abstract. This paper presents an application of adaptive neural networks based on a dynamic neural network to trajectory tracking of unknown nonlinear plants. The main methodologies on which the approach is based are recurrent neural networks and Lyapunov function methodology and Proportional-Integral (PI) control for nonlinear systems. The proposed controller structure is composed of a neural identifier and a control law defined by using the PI approach. The new control scheme is applied via simulations to Chaos Synchronization. Experimental results have shown the usefulness of the proposed approach for Chaos Production. To verify the analytical results, an example of a dynamical network is simulated and a theorem is proposed to ensure tracking of the nonlinear system.

Keywords. Dynamic neural networks, chaos production, chaos synchronization, trajectory tracking, Lyapunov function stability, PI control.

1 Introduction

Artificial neural networks as computational models of the brain are widely used in engineering applications due to their ability to estimate the relation between inputs and outputs from a learning process. Motivated by the seminal paper [1], there exists a continuously increasing interest in applying neural networks to identification and control of nonlinear systems. Most of these applications use feedforward structures [2, 3]. Recently, recurrent neural networks were developed as an extension of the static neural network capability to approximate nonlinear functions, therefore recurrent neural networks can approximate nonlinear systems. They allow more efficient modeling of the underlying dynamical systems [4].

Three representative books [5, 6, 7] reviewed the application of recurrent neural networks to nonlinear system identification and control. In particular, [5] uses off-line learning, while [6] analyzes adaptive identification and control by means of on-line learning, where stability of the closed-loop system is established based on the Lyapunov function method. In [6], the trajectory tracking problem is reduced to a linear model following problem, with application to DC electric motors. In [7], analyses of Recurrent Neural Networks for identification, estimation and control are developed, with applications.
to chaos control, robotics and chemical processes. Main approaches include the use of differential geometry theory [8]. Recently, the passivity approach has generated increasing interest for synthesizing control laws [9]. An important problem for these approaches is how to achieve robust nonlinear control in the presence of unmodelled dynamics and external disturbances. In this direction, there exists the so-called $H_{sc}$ nonlinear control approach [10].

One major difficulty with this approach, alongside with its possible system structural instability, seems to be the requirement of solving some resulting partial differential equations. In order to alleviate this computational problem, the so-called inverse optimal control technique was recently developed, based on the input-to-state stability concept [11]. On the basis of the inverse optimal control approach, a control law for generating chaos in a recurrent neural network was designed in [12]. In [13, 14] this methodology was modified for stabilization and trajectory tracking of an unknown chaotic dynamical system. The proposed adaptive control scheme is composed of a recurrent neural identifier and a controller, where the former is used to build an on-line model for the unknown plant and the latter, to ensure the unknown plant to track the reference trajectory. In this paper, we further improve the design by adequating it to systems with less inputs than states. The approach is based on the methodology developed in [13, 14], in which the control law is optimal with respect to a well-defined Lyapunov function.

The new approach is illustrated by Chaos Synchronization as an example of a complex dynamical system.

2 Modeling of the Plant

The unknown nonlinear plant is given by

$$\dot{x}_p = F_p(x_p, u) \triangleq f_p(x_p) + g_p(x_p)u,$$  \hspace{1cm} (1)

where $x_p, f_p \in \mathbb{R}^n, u \in \mathbb{R}^m$ and $g_p \in \mathbb{R}^{n \times m}$. $x_p$ is the plant, $u$ is the control input and both $f_p$ and $g_p$ are unknown. We propose to model (1) by the neural network state space representation

$$\dot{\tilde{x}}_p = A\tilde{x}_p + W^*\Gamma_z(x) + \Omega u,$$  \hspace{1cm} (2)

where $\tilde{x}_p = x + w_{per}$, where $x \in \mathbb{R}^n$ is the state variable of the unknown plant, $w_{per}$ is the modeling error between the neural network and the plant given by

$$w_{per} = x - \tilde{x}_p.$$ \hspace{1cm} (3)

We define the modeling error between the neural network and the plant by

$$w_{per} = x - \tilde{x}_p.$$ \hspace{1cm} (4)

We assume the following.

Hypotheses 1. (Objective of Modeling): Modeling error is exponentially stable, that is,

$$w_{per} \to 0 \text{ when } t \to \infty.$$ \hspace{1cm} (5)

This condition guarantees that $w_{per} \to 0$ when $t \to \infty$ and it is usually found in dynamical systems that model real applications.

In this work we consider $k = 1$, and now, from (2) we have $\dot{w}_{per} = \lambda w_{per}$, where $\dot{w}_{per} = x + w_{per}$.

The unknown plant can be modeled as

$$\dot{x}_p = \dot{x} + w_{per} = A(x) + W^*\Gamma_z(x) + \Omega u.$$ \hspace{1cm} (6)

3 Trajectory Tracking

We proceed now to analyze the modeling error between the unknown plant modeled by (4) and the reference signal defined by

$$\dot{x}_r = f_r(x_r, u_r), \text{ with } u_r \text{ and } x_r \in \mathbb{R}^n,$$ \hspace{1cm} (7)

where $x_r$ are the reference states, $u_r$ is the input and $f_r$ is a nonlinear function.

For this purpose, we define the control error between the plant and the reference signal by

$$e = x_p - x_r,$$ \hspace{1cm} (8)

whose derivative with respect to time is

$$\dot{e} = \dot{x}_p - \dot{x}_r = A(x) + W^*\Gamma_z(x) + w_{per}.$$ \hspace{1cm} (9)

The error term.

A the neural network state space representation $u$ the plant, $x = g(x) + \Gamma(z(x)) + \Omega u$, where $\Gamma(z(x))$ is the so-called external disturbances. In this direction, there exists the so-called $H_{sc}$ nonlinear control approach [10].

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Adding and subtracting to the right hand side from (7) the terms \( \hat{W} T_z(x_r) + \alpha_r(t, \hat{W}) \), \( A \hat{e} \), where \( \hat{W} \) is the estimate of \( W^* \) and taking into account that \( \hat{w}_p = x - x_p \), we have

\[
\begin{align*}
\dot{e} &= A(x) + W^* T_z(x) + x - x_p + \Omega u - f_r(x_r, u_r) \\
&+ \hat{W} T_z(x_r) - \hat{W} T_z(x_r) + \Omega \alpha_r(t, \hat{W}) \\
&- \Omega \alpha_r(t, \hat{W}) - e - x_r - A e + x + A(x).
\end{align*}
\]

Then

\[
\Omega \alpha_r(t, \hat{W}) = f_r(x_r, u_r) - A x_r - \hat{W} T_z(x_r) - x_r + x_p
\]

and we get

\[
\dot{e} = A e + W^* T_z(x) - \hat{W} T_z(x_r) - A e \\
+ (A + I)(x - x_r) \\
+ \Omega (u - \alpha_r(t, \hat{W})).
\]

Now, adding and subtracting in (10) the term \( \hat{W} \) \( \Gamma_z(x) \) and defining \( \Gamma_z(x) = \Gamma(z(x) - z(x_r)) \) we have

\[
\begin{align*}
e &= A e + (W^* - \hat{W}) \Gamma_z(x) + \hat{W} T_z(x) - z(x_r) \\
&+ (A + I)(x - x_r) - A e + \Omega (u - \alpha_r(t, \hat{W})).
\end{align*}
\]

We define

\[
\hat{W} = W^* - \hat{W} \text{ and } \hat{u} = u - \alpha_r(t, \hat{W})
\]

and replacing (12) in (11), we obtain

\[
\begin{align*}
\dot{e} &= A e + \hat{W} T_z(x) + \hat{W} T_z(z(x) - z(x_r)) \\
&+ (A + I)(x - x_r) - A e + \hat{W} \tilde{u},
\end{align*}
\]

\[
\begin{align*}
\dot{e} &= A e + \hat{W} T_z(x) + \hat{W} T(z(x) - z(x_p) + z(x_p)) \\
&- z(x_r)) + (A + I)(x - x_p + x_p - x_r) \\
&- A e + \hat{W} \tilde{u}.
\end{align*}
\]

Now we set

\[
\tilde{u} = u_1 + u_2.
\]

So we define

\[
\Omega u_1 = -\hat{W} T(z(x) - z(x_p)) - (A + I)(x - x_p)
\]

and (13) is reduced to

\[
\begin{align*}
\dot{e} &= A e + \hat{W} T_z(x) + \hat{W} T(z(x) - z(x_r)) \\
&+ (A + I)(x - x_r) - A e + \Omega u_2.
\end{align*}
\]

Considering that \( e = x_p - x_r \), the last equation can be written as

\[
\dot{e} = (A + I)e + \hat{W} T_z(x) + \hat{W} T(z(e + x_r) - z(x_r)) + \Omega u_2,
\]

If \( \phi(e) = \sigma(e + x_r) - \sigma(x_r) \), we get

\[
\dot{e} = (A + I)e + \hat{W} \sigma(x) + \hat{W} (\sigma(e + x_r) - \sigma(x_r)) + \Omega u_2.
\]

Now the problem is to find the control law \( \Omega u_2 \) that stabilizes the system (16). We will obtain the control law by using the Lyapunov methodology.
4 Stability of the Tracking Error

Once (16) is obtained, we consider its stabilization in feedforward networks. We note that \((e, \tilde{W}) = 0\) is an asymptotically stable equilibrium point of the undisturbed autonomous system \((A = -\lambda I \text{ and } \lambda > 0)\). For its stability, we propose the next PI control law:

\[
\Omega u_2 = K_pe + K_i \int_0^t e(\tau)d\tau - Y(\frac{1}{2} + \frac{1}{2}||\tilde{W}||^2 L_\phi^2)e. \tag{17}
\]

The parameters \(K_p\) and \(K_i\) will be determined later, and \(L_\phi^2\) is the Lipschitz constant of \(\phi_x\), with \(Y > 0\).

We will show that the feedback system is asymptotically stable. Replacing (17) in (16), we obtain

\[
\dot{e} = (A + I)e + \tilde{W}\sigma(x) + \tilde{W}\phi(e) + K_pe + K_i \int_0^t e(\tau)d\tau \tag{18}
\]

\[
= -(\lambda - 1 - K_p)e + \tilde{W}\sigma(x) + \tilde{W}\phi(e) + K_i \int_0^t e(\tau)d\tau \tag{19}
\]

and if \(w = K_i \int_0^t e(\tau)d\tau\), then \(\dot{w} = K_i e(\tau)d\tau\), so we can rewrite (19) as

\[
\dot{e} = -(\lambda - 1 - K_p)e + \tilde{W}\sigma(x) + \tilde{W}\phi(e) + \dot{w} \tag{20}
\]

We will show that the new state \((e, w)^T\) is asymptotically stable and that the equilibrium point is \((e, w)^T = (0, 0)^T\), when \(\tilde{W}\sigma(x_r) = 0\), which is taken as an external disturbance.

Let \(V\) be the candidate Lyapunov function given by

\[
V = \frac{1}{2}(e^T, w^T)(e, w)^T + \frac{1}{2} tr \left\{ \tilde{W}^T \tilde{W} \right\}. \tag{21}
\]

The time derivative of (21) along the trajectories of (20) is

\[
\dot{V} = (e^T, w^T)(\dot{e}, \dot{w})^T + tr \left\{ \tilde{W}^T \tilde{W} \right\}
\]

\[
= e^T \dot{e} + w^T \dot{w} + tr \left\{ \tilde{W}^T \tilde{W} \right\}. \tag{22}
\]

\[
\dot{V} = e^T (\lambda - 1 - K_p)e + \tilde{W}\sigma(x) + \tilde{W}\phi(e) + w^T K_i e - \gamma(\frac{1}{2} + \frac{1}{2}||\tilde{W}||^2 L_\phi^2))e \\
+ \gamma(\frac{1}{2} + \frac{1}{2}||\tilde{W}||^2 L_\phi^2) e \\
+ tr \left\{ \tilde{W}^T \tilde{W} \right\}. \tag{23}
\]

In this part, we select the following learning law from the neural network weights as in [6] and [15]:

\[
tr \left\{ \tilde{W}^T \tilde{W} \right\} = -e^T \tilde{W}\sigma(x). \tag{24}
\]

Then (23) is reduced to

\[
\dot{V} = (\lambda - 1 - K_p)e^T e + e^T \tilde{W}\phi(e) + (1 + K_i)e^T w \\
+ \gamma(\frac{1}{2} + \frac{1}{2}||\tilde{W}||^2 L_\phi^2) e^T e. \tag{25}
\]

We apply the next inequality to the second term in the right hand side of (25)

\[
x^T y \leq \frac{1}{2} x^T x + \frac{1}{2} y^T y \tag{26}
\]
to get
\[
\dot{V} \leq -(\lambda - 1 - K_p)e^T e + \left(1 + \frac{1}{2}\right)\|\hat{W}\|^2 L_{p}^2 e^T e + (1 + K_i)e^T w \\
-\gamma\left(1 + \frac{1}{2}\right)\|\hat{W}\|^2 L_{p}^2 e^T e.
\]  
(27)

The parameters in (27) are reduced to
\[
\dot{V} \leq -(\lambda - 1 - K_p)e^T e - (\Psi - 1)(1 + \frac{1}{2}\|\hat{W}\|^2 L_{p}^2 e^T e).
\]  
(28)

Here, if we choose \(\lambda - 1 - K_p > 0\) and \(\Psi - 1 > 0\), then \(V < 0\), \(\forall e, w, \hat{W} \neq 0\), the error tracking is asymptotically stable and it converges to zero for every \(e \neq 0\), this means that the plant follows the reference asymptotically. Finally, the control law which affects the plant and the neural network is given by
\[
u = \Omega^\dagger[-\hat{W}^T(z(x) - z(x_p)) - (A + I)(x - x_p) + K_p e + K_i \int_0^t e(\tau)d\tau - \Psi(1 + \frac{1}{2}\|\hat{W}\|^2 L_{p}^2 e) + f_r(x_r, u_r) - Ax_r - \hat{W}z(x_r) - x_r + x_p]
\]  
(29)

**Remark 1**: \(\Omega^\dagger\) is the pseudo inverse in the sense of Moore–Penrose.

This control law gives asymptotic stability of error dynamics and thus ensures the tracking to the reference signal. The results obtained can be summarized as follows.

**Theorem 2**: For the unknown nonlinear system (1) modeled by (4), the on-line learning law (24) and the control law (29) together ensure the tracking to the nonlinear reference model (5).

**Remark 3**: From (28) we have
\[
\dot{V} \leq -(\lambda - 1 - K_p)e^T e - (\Psi - 1)(1 + \frac{1}{2}\|\hat{W}\|^2 L_{p}^2 e^T e < 0, \forall e \neq 0, \forall \hat{W}.
\]

where \(V\) is decreasing and bounded from below by \(V(0)\), and then \(V = \frac{1}{2}(e^T, w^T)(e, w)^T + \frac{1}{2}tr\left\{\hat{W}^T \hat{W}\right\}\.

We conclude that \(e, \hat{W} \in L_1\); this means that the weights remain bounded.

### 5 Simulations

In order to demonstrate the applicability of the proposed adaptive control scheme, the following example is tested.

In this example, the unknown plant considered is Chen’s chaotic attractor generated by
\[
\begin{align*}
x_{p1} &= 35x_{p2} - 35x_{p1}, \quad x_{p1}(0) = -10, \\
x_{p2} &= -7x_{p1} - x_{p1}x_{p3} + 28x_{p2}, \quad x_{p2}(0) = 0, \\
x_{p3} &= x_{p1}x_{p2} - 3x_{p3}, \quad x_{p3}(0) = 37. 
\end{align*}
\]  
(30)

The goal is to force the chaotic Chen’s attractor to track the reference—the Roosler attractor—generated by
\[
\begin{align*}
x_{r1} &= -x_{r2} - x_{r3}, \quad x_{r1}(0) = 0.1, \\
x_{r2} &= x_{r1} + 0.2x_{r2}, \quad x_{r2}(0) = 0, \\
x_{r3} &= x_{r1}x_{r3} - 5.7x_{r3} + 0.2, \quad x_{r3}(0) = 0. 
\end{align*}
\]  
(31)

In the simulations, the following dynamic neural network was used:
\[
\dot{x} = Ax + W^* \Gamma z(x) + \Omega u 
\]  
(32)

with
\[
A = \begin{pmatrix}
-5 & 0 & 0 \\
0 & -5 & 0 \\
0 & 0 & -5
\end{pmatrix}
\]
and
\[
\Gamma = \begin{pmatrix}
150 & 0 & 0 \\
0 & 150 & 0 \\
0 & 0 & 150
\end{pmatrix},
\]
while
\[
z(x) = \begin{pmatrix}
tanh(0.5x_1) \\
tanh(0.5x_2) \\
tanh(0.5x_3)
\end{pmatrix}.
\]
$W^*$ is estimated using the learning law given in (24), and the $u$ is calculated using (29).

The results of simulations are shown in Figures 1-11, where the time evolution of the states and phase portraits are presented.

We can see that the Recurrent Neural Controller ensures rapid convergence of the system outputs to the reference trajectory [17]. Another important issue of this approach related to other neural controllers is that most neural controllers are based on indirect control: firstly, the neural network identifies the unknown system and when the identification error is small enough, the control is applied. In our approach, direct control is considered, the learning laws for the neural networks depend explicitly on the tracking error instead of the identification error.

This approach results in faster response of the system.

6 Conclusions

We have extended the adaptive recurrent neural control previously developed in [13,14,16] to trajectory tracking control problem in order to consider less inputs than states. Stability of the tracking error is analyzed via Lyapunov control functions and the control law is obtained based on the PI approach. A new adaptive control structure based on a dynamic neural network for chaotic orbit tracking of unknown nonlinear systems has been developed. This structure is composed of a neural network identifier and a control law for orbit tracking. Stability of the tracking control system has
The applicability of the proposed structure was tested via simulations, using Chaos Synchronization as an example of a complex dynamical system. The results are quite encouraging. Research along this line will consist in further relaxing the required condition of having the same number of inputs and states in the control system. Also, we will continue to implement the control algorithm in real time and to perform more tests in a laboratory.

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References


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